

# Wireless Power Transfer to Mobile Wearable Device via Resonance Magnetic

Olutola Jonah and Stavros V. Georgakopoulos  
 Department of Electrical Engineering,  
 Florida International University, Miami, FL 33174,  
 USA

Manos M. Tentzeris  
 The School of Electrical and Computer Engineering,  
 Georgia Institute of Technology, Atlanta,  
 GA , 30332, USA

**Abstract**—The wireless power transfer (WPT) efficiency of a mobile wearable medical device (MWMD) by Strongly Coupled Magnetic Resonance (SCMR) method is studied here for air-tissue interfaces. Specifically, SCMR’s wireless power transfer from a source of air to a sensor close to tissue is analyzed. A comparison of wireless power transfer efficiency between resonant spirals at different distances are shown. The efficiencies achieved in different areas of the human body are also reported

**Keywords**- Wireless, magnetic resonance, wearable devices, tissue.

## I. INTRODUCTION

Wearable device application has an increasing variety of applications, this is used to monitor the health status of elderly patients or patients undergoing medical care at home, which can reduce cost and improve comfort [1]. It is also used for tracking human body kinematics to allow clinicians to classify and analyze a stroke patient’s progress, which aid in the rehabilitation plan [2]. A mobile wearable medical device (MWMD) is also used in high-risk cardiac and respiratory patients to monitor and alert as necessary, the system might include periodical or continuous collection and evaluation of multiple vital signs [3]. MWMDs are saving and extending lives, due to their ability to monitor, stimulate and regulate vital organs, and also communicate with host about the state of health of these organs intelligently.

Wireless power transmission system for biomedical applications requires strict power level, in order to avoid excessive heating of the tissue and resulting in significant tissue damage [4]. This restriction requires the design of low power source with high transmission efficiency for such system [5]. Three popular wireless powering techniques, which were proposed for the air-to-air transmission, have also used for the air-tissue scheme to meet the air-tissue requirement mentioned above: (a) inductive coupling, (b) strongly coupled magnetic resonance and (c) electromagnetic radiation. The strongly coupled magnetic resonance (SCMR) is the most promising of the three methods for MWMDs because of its high efficiency and range.

The SCMR method employs resonators to transmit power wirelessly and efficiently over mid-range distances [6], where the adverse effects of the low coupling coefficient

between the two coils for inductive coupling are compensated for by the high-Quality factor of the four-element system to achieve high efficiency. The SCMR method is a non-radiative wireless mid-range power transfer method (10 – 300 cm) that has been recently developed [7–10]. It achieved approximately 40% efficiency in the air at a distance of 2 m with a single receiver [7]. In addition, the technique was also used to simultaneously power multiple receivers in the air, and approximately 60% efficiency was attained at a distance of 2 m in the air [10]. In [11] this technique was extended to the in-vitro and in-vivo experiments.

## II. WPT IN SCMR WITH SPIRALS

The SCMR systems use resonant transmitters and receivers that are strongly coupled. Strongly coupled systems are able to transfer energy efficiently, because resonant objects exchange energy efficiently versus non-resonant objects that only interact weakly [7]. A standard SCMR system consists of four elements (typically four loops, or two loops and two coils). Here, an SCMR system based on spirals is shown in Fig. 1.

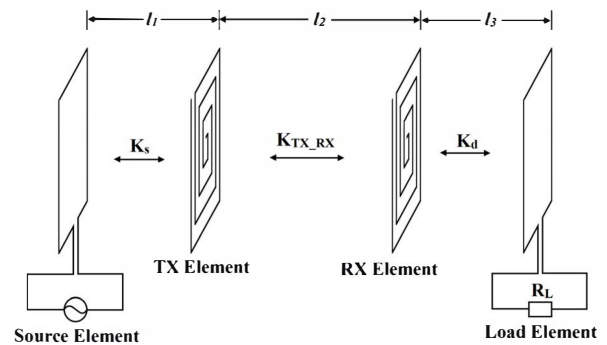


Fig. 1. Schematic of an SCMR system with spirals in the air, where  $K_s$ ,  $K_{TX\_RX}$  and  $K_d$  are the respective coupling coefficients.

The source element is connected to the power source, and it is inductively coupled to the TX element. The TX element must exhibit a natural resonance frequency that is identical to the RX. Both elements should be resonant at the frequency, where their Q-factor is naturally maximum. Furthermore, the load element is terminated with a load. For our analysis, we assume that the entire system operates in air. The resonance frequency of the spiral,  $f_r$ , can be

calculated from [4]:

$$f_r = \frac{1}{2\pi\sqrt{LC}} \quad (1)$$

The resonant frequency,  $f_r$  is also the operational frequency of the SCMR wireless powering system. The Q-factor of the resonance frequency can be written as:

$$Q = \frac{2\pi f_r L}{R_{\text{ohm}} + R_{\text{rad}}} \quad (2)$$

where,  $L$ ,  $R_{\text{ohm}}$ , and  $R_{\text{rad}}$  are the self-inductance, ohmic resistance and radiation resistance of the spiral. The inductance,  $L$ , of a spiral can be written as [13]:

$$L = \left[ \frac{\mu_0 N^2 \left( \frac{d_{\text{in}} + d_{\text{out}}}{2} \right) c_1}{2} \right] \left[ \ln \left( \frac{c_2}{\alpha} \right) + c_3 \alpha + c_4 \alpha^2 \right] \quad (3)$$

where,  $c_1 = 1.27$ ,  $c_2 = 2.07$ ,  $c_3 = 0.18$  and  $c_4 = 0.13$ , are the constants derived based on the geometrical layout of the square spiral; and  $\alpha$  is the fill ratio defined by  $\alpha = (d_{\text{in}} - d_{\text{out}})/(d_{\text{in}} + d_{\text{out}})$ . The ohmic and radiation resistances can be written as [11], [12]:

$$R_{\text{ohm}} = \frac{\ell_{\text{tot}}}{4\sqrt{WT}} \sqrt{\pi\mu_0\rho f} \left( 1 + \frac{R_p}{R_o} \right) \quad (4)$$

$$R_{\text{rad}} = 31200 \left( \frac{f}{c} \right)^4 \left( \sum_{i=1}^N d_i^2 \right)^2 \quad (5)$$

where,  $d_i$  is the side length of the  $i$ th turn of the spiral,  $\rho$  is the spiral's conductor resistivity,  $c$  is the speed of light, and  $\sqrt{\pi\mu_0\rho f}$  represents the conductor's sheet resistance [12]. The factor  $R_p/R_o$  in (4) represents the proximity effect factor that accounts for the additional resistance due to the closeness of the conductors. In order to derive analytical expressions for  $Q_{\text{max}}$  and  $f_{\text{max}}$ , the analytical and simulation setups are chosen such that the proximity effect is negligible reducing (4) to:

$$R_{\text{ohm}} = \frac{\ell_{\text{tot}}}{4\sqrt{WT}} \sqrt{\pi\mu_0\rho f} \quad (6)$$

It should also be noted that (3) – (6) are effective in SCMR analysis only when  $\ell_{\text{tot}} < \lambda/3$  [12]. The Q-factor of a resonant spiral can be expressed in terms of its geometrical parameters using (2), (4), (5) and (6) [13] as:

$$Q = \frac{\pi f_r \mu_0 N^2 \left( \frac{d_{\text{in}} + d_{\text{out}}}{2} \right) c_1 \left[ \ln \left( \frac{c_2}{\alpha} \right) + c_3 \alpha + c_4 \alpha^2 \right]}{\frac{\ell_{\text{tot}}}{4\sqrt{WT}} \sqrt{\pi\mu_0\rho f_r} + 31200 \left( \frac{f_r}{c} \right)^4 \left( \sum_{i=1}^N d_i^2 \right)^2} \quad (7)$$

The maximum possible Q-factor,  $Q_{\text{max}}$ , of a spiral and the frequency,  $f_{\text{max}}$ , where  $Q_{\text{max}}$  occurs, can be derived from (7) using standard calculus as:

$$f_{\text{max}} = 120.44 \times 10^6 \left[ \frac{\ell_{\text{tot}} \sqrt{\mu_0 \rho}}{\sqrt{WT} \left( \sum_{i=1}^N d_i^2 \right)^2} \right]^{2/7} \quad (8)$$

$$Q_{\text{max}} = \frac{\pi f_{\text{max}} \mu_0 N^2 \left( \frac{d_{\text{in}} + d_{\text{out}}}{2} \right) c_1 \left[ \ln \left( \frac{c_2}{\alpha} \right) + c_3 \alpha + c_4 \alpha^2 \right]}{4\sqrt{WT} \sqrt{\pi\mu_0\rho f_{\text{max}}} + 31200 \left( \frac{f_{\text{max}}}{c} \right)^4 \left( \sum_{i=1}^N d_i^2 \right)^2} \quad (9)$$

### III. WIRELESS POWERING OF MWMD DEVICE

Equations (8) and (10) are used to calculate the geometrical parameters of the TX and RX spirals in order to achieve maximum power transfer and minimize tissue exposure to electromagnetic field. This is based on the FCC specification the exposure limits for tissue to radio frequency (RF) energy from wireless devices for biomedical applications [14]. The specification of the models are: The TX spiral model dimensions are  $N = 45$ ,  $d_{\text{in}} = 4\text{mm}$   $d_{\text{out}} = 272\text{mm}$ ,  $w = 2\text{mm}$  and the material is copper. The RX spiral model dimensions are  $d_{\text{in}} = 3.8\text{mm}$   $d_{\text{out}} = 243\text{mm}$ ,  $w = 1.7\text{mm}$ . The distance between TX and RX Spiral are 10, 20 and 30 cm. The RX spiral is 3 mm from the skin and embedded in a dielectric with a permittivity value of 4.2 to simulate clothing effect. The models were simulated in HFSS with Human phantom model in which the various tissues are already characterized. Fig. 2 shows the part of the human phantom included in our simulations with the planar SCMR system. The efficiency plots for the different parts of the body examined are shown in Fig. 3 and Fig. 4.

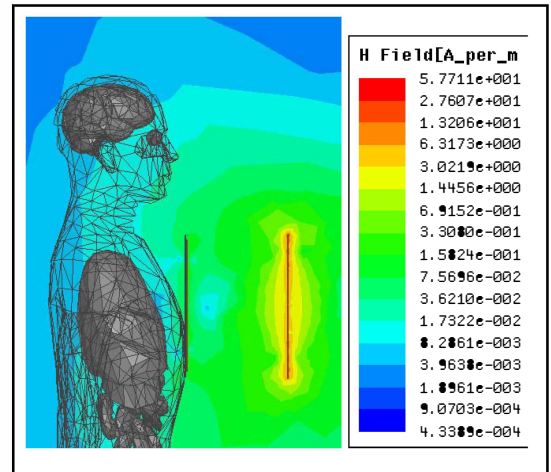


Figure. 2. The human body and magnetic field distribution at  $l_2 = 20\text{ cm}$ .

The result of the SAR, magnetic and electric field distribution in two different parts of the body examined are shown in tables I and II.

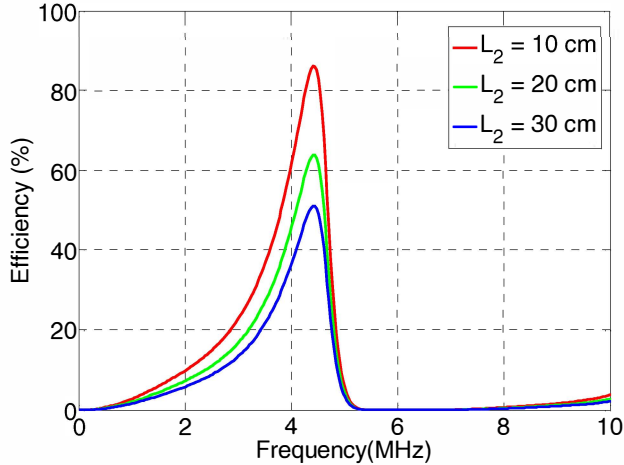


Figure 3. Efficiency plot human chest at 3 mm distance.

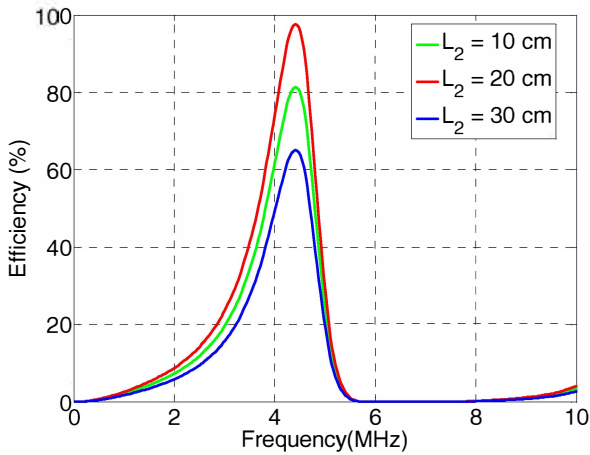


Figure 4. Efficiency plot human head at 3 mm distance.

TABLE I  
COMPARISON OF DIFFERENT DEPTH  
WITH INPUT POWER OF 1 WATT IN FRONT OF THE CHEST

Distance (cm)	Field Parameters and Efficiency			
	Max E-field (V/m)	Max H-field (A/m)	Max SAR (W/kg)	Efficiency (%)
10	18	0.21	0.07	83
20	12	0.08	0.0004	64
30	0.9	0.004	0.00008	56

TABLE II  
COMPARISON OF DIFFERENT DIFFERENT DEPTH  
WITH INPUT POWER OF 1 WATT BESIDE THE HEAD

Distance (cm)	Field Parameters and Efficiency			
	Max E-field (V/m)	Max H-field (A/m)	Max SAR (W/kg)	Efficiency (%)
10	15	0.15	0.0005	92
20	5	0.05	0.00001	81
30	1.2	0.001	0.000002	63.5

The result shows that the SCMR has great potentials in MWMD, and has very low SAR, electric and magnetic field distribution close to the tissue, which is important for biomedical applications. It is planner, hence can be embedded in clothing easily.

#### ACKNOWLEDGMENT

This work was supported in part by the Dissertation Year Fellowship that was provided by Florida International University.

#### REFERENCES

- [1] E. Sardini, M. Serpelloni, Instrumented wearable belt for wireless health monitoring, *Procedia Engineering*, Volume 5, 2010, Pages 580-583, ISSN 1877-7058, 10.1016/j.proeng.2010.09.176
- [2] Chee Kian Lim, Zhiqiang Luo, I-Ming Chen, Song Huat Yeo, Wearable wireless sensing system for capturing human arm motion, *Sensors and Actuators A: Physical*, Volume 166, Issue 1, March 2011, Pages 125-132, ISSN 0924-4247, 10.1016/j.sna.2010.10.015.
- [3] P.S. Pandian, K. Mohanavelu, K.P. Safeer, T.M. Kotresh, D.T. Shakunthala, Parvati Gopal, V.C. Padaki, Smart Vest: Wearable multi-parameter remote physiological monitoring system, *Medical Engineering & Physics*, Volume 30, Issue 4, May 2008, Pages 466-477, ISSN 1350-4533, 10.1016/j.medengphy.2007.05.014.
- [4] M. Kiani and M. Ghovanloo, "An RFID-Based Closed-Loop Wireless Power Transmission System for Biomedical Applications," *IEEE Transactions on Circuits and Systems-II: EXPRESS BRIEFS*, Vol. 57, No.4, pp. 260-265, April 2010.
- [5] A. Ramrakhiani, S. Mirabbasi and M. Chiao, "Design and Optimization of Resonance-Based Efficient Wireless Power Delivery Systems for Biomedical Implants," *IEEE Transactions on Biomedical Circuits and Systems*, Vol. 5, No. 1, pp. 48-63, February 2011
- [6] A. Kurs, A. Karalis, R. Moffatt, J. D. Joannopoulos, P. Fisher, and M. Soljagic, "Wireless Power Transfer via Strongly Coupled Magnetic Resonances," *Science*, Vol. 317, No.5834, pp. 83-86, July 2007.
- [7] A. Karalis, J.D. Joannopoulos, and M. Soljagic, "Efficient wireless non-radiative mid-range energy transfer," *Elsevier, Annals of Physics*, vol. 323, pp. 34-48, January 2008.
- [8] A. Kurs, A. Karalis, R. Moffatt, J. D. Joannopoulos, P. Fisher, and M. Soljagic, "Wireless Energy Transfer via Strongly Coupled Magnetic Resonances," *Science*, vol. 317, pp. 83-85, 2007.
- [9] D. Joannopoulos, A. Karalis, and M. Soljagic, "Wireless Non-Radiative energy transfer," *US Patent 20070222542*, Sept. 2007.
- [10] A. Kurs, A. Karalis, R. Moffatt and M. Soljagic Marin, simultaneous midrange power transfer to multiple devices, *Applied Physics Letter*, vol. 96, 044102, 2010.
- [11] F. Zhang, X. Liu, S. A. Hackworth, R. J. Scwabassi and M. Sun, "In Vitro and In Vivo Studies on Wireless Powering of Medical Sensors and Implantable Devices," *IEEE/NIH Life Science Systems and Applications Workshop*, pp.84-87, 2009.
- [12] C.A. Balanis, *Antenna Theory: Analysis and Design*, Wiley, New Jersey, 2005, ch. 5.
- [13] O. Jonah, A Merwaday, S. V. Georgakopoulos and Manos M. Tentzeris, "Spiral Resonators for Optimally Efficient Strongly Coupled Magnetic Resonant Systems," unpublished.
- [14] D. Poljak, *Human Exposure to Electromagnetic Fields*, WIT Press, LLC, ISBN 1-85312-997-6, 2004.